2D BEAMFORMING ON SPARSE ARRAYS WITH SPARSE BAYESIAN LEARNING

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2D BEAMFORMING ON SPARSE ARRAYS WITH SPARSE BAYESIAN LEARNING

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ABSTRACT
Sparse arrays such as co-prime and nested arrays can identify more sources than the number of sensors. This is because their difference co-arrays contain a uniformly spaced virtual array with more elements than the number of sensors in the array. In this paper we demonstrate this using two-dimensional co-prime and nested sparse arrays combined with sparse Bayesian learning (SBL) for 2D beamforming in azimuth and elevation. SBL can directly process the sparse array data and significantly outperform conventional beamforming and MUSIC as seen from simulations.

Index Terms—2D beamforming, SBL, co-prime arrays, nested arrays, compressive sensing

1. INTRODUCTION
A two-dimensional (2D) planar array can estimate the azimuth and elevation directions (2D beamforming) of plane waves incident on it. The uniform rectangular array (URA) often used for this purpose is too dense and requires large number of sensors to achieve a given resolution. Sparse arrays can significantly reduce the number of sensors and resolve more source directions than the number of sensors when the incident signals are spatially wide sense stationary.

Various sparse array geometries have been proposed in the literature, here we focus on co-prime and nested arrays. In one-dimension, linear co-prime [1] and nested [2] arrays are well studied [3, 4, 5, 6, 7, 8]. More recently, 2D nested arrays [9, 10] and multi-dimensional co-prime arrays [11] have been proposed and used for 2D direction-of-arrival (DOA) estimation [12, 13, 14]. In this paper we use the terms beamforming and DOA estimation synonymously.

Sparse Bayesian learning (SBL) [15, 16] is a sparse processing method that has been successfully used for beamforming with uniform linear arrays [17, 18, 19, 20, 21] and 1D sparse arrays [7, 8, 22]. SBL formulates the beamforming problem probabilistically and performs stochastic maximum likelihood to estimate the source covariance parameter. Sparsity profile of the source covariance diagonal corresponds to the source direction-of-arrivals (DOAs). Our previous work [7, 8] explored SBL for processing measurements from 1D sparse arrays. In this paper we extend our analysis for 2D sparse arrays. We give examples of 2D sparse arrays and use them to beamform source azimuth and elevation parameters. The sparse arrays also identify more sources than the number of sensors.

2. 2D SPARSE ARRAYS
In this paper we are concerned with 2D planar arrays. A 2D uniform rectangular array (URA) has sensors located on a uniform grid with spacing \( d = \lambda/2 \) along both directions. Here \( d \) is the fundamental grid spacing. For a given aperture a URA has a dense spacing of sensors to avoid aliasing.

Sparse arrays are an alternative to URA which use relatively fewer sensors. When the incident signals are wide sense stationary, sparse arrays can capture all the degrees of freedom captured by the corresponding full URA. There are many classes of sparse arrays, two of which are used in this paper, co-prime and nested arrays, and are briefly discussed next. A detailed mathematical analysis of the 2D co-prime [11] and nested [9, 10] arrays is beyond the scope of this paper.

2.1. Co-prime arrays
2D co-prime arrays are constructed by combining two URAs each of which are generated using a \( 2 \times 2 \) sampling matrix. These sampling matrices are relatively co-prime and their product commutes (see [11] for definitions and details). An example of a 2D co-prime array is shown in Fig. 1(a) and its difference co-array in Fig. 1(c). The array consists of a 9-sensor (red dots) URA with \( 2 \times \lambda/2 \) spacing in both directions and a 16-sensor (blue dots) URA with spacing \( 3 \times \lambda/2 \). The difference co-array (or simply co-array) is obtained by computing the pairwise differences between all the sensor positions. The number of elements (239) in the co-array is indicative of the degrees of freedom resolvable by the array.

2.2. Nested arrays
A two dimensional nested array consists of two URAs: a dense URA and a sparse URA. A 2D nested array example is shown in Fig. 1(b) along with its difference co-array in Fig. 1(d). The array consists of a dense 9-sensor (red dots) URA with \( 1 \times \lambda/2 \) spacing in both directions and a less dense
16-sensor (blue dots) URA with spacing $3 \times \lambda/2$. The difference co-array of a nested array has no holes while that of a co-prime array has holes which is seen in Fig. 1. The number of elements (263) in the nested-array is indicative of the degrees of freedom resolvable by the array, which is similar to that of the co-prime array. Though the examples of sparse arrays considered here lie on rectangular grids, the analysis we develop is valid for other array realizations as well.

$$\gamma_m^{\text{old}} = a_m^H S_y a_m, \forall m$$

4. for $i = 1$ to $N_i$
5. Compute: $\Sigma_y = \sigma^2 I_N + A \Gamma^{\text{old}} A^H$
6. $\gamma_m^{\text{new}}$ update $\forall m$ using (8)
7. If $||\gamma^{\text{new}} - \gamma^{\text{old}}||_1 < \epsilon$, break
8. $\gamma^{\text{old}} = \gamma^{\text{new}}, \Gamma^{\text{old}} = \text{diag}(\gamma^{\text{new}})$
9. end
10. Output: $\gamma$

**Table 1.** SBL algorithm pseudocode: Input consists of data matrix $Y$, dictionary $A$, and noise variance $\sigma^2$. Convergence is controlled by the error threshold $\epsilon$ and maximum number of iterations $N_i$.

Let $(\theta_1, \phi_1), \ldots, (\theta_M, \phi_M)$ be a discrete grid in the $\theta - \phi$ plane. Assume that this grid is sufficiently fine so that the true source directions are among these $M$ candidate directions. Let $A = [a_1, \ldots, a_M]$ be the dictionary of steering vectors with $a_k$ corresponding to direction $(\theta_k, \phi_k)$. The observation $y$ is expressed as

$$y = Ax + n,$$  \hspace{1cm} (3)$$

where $x \in \mathbb{C}^M$ has exactly $K$ non-zero entries at locations corresponding to the $K$ source directions and having value of source amplitudes. Typically $K \ll M$ which makes (3) an underdetermined system which can be solved using compressive sensing methods. In this paper we use SBL to solve for the covariance of the complex source amplitudes $x$ in (3).

We briefly discuss the Sparse Bayesian learning algorithm. For more detailed description readers can refer to any of the numerous references [15, 16, 18, 21]. When multiple snapshots are available, (3) can be written as

$$Y = AX + N,$$ \hspace{1cm} (4)$$

where the noise $N = [n_1, \ldots, n_L]$ is zero-mean complex Gaussian with variance $\sigma^2$, $n_l \sim \mathcal{CN}(n_l; 0, \sigma^2 I)$; $X = [x_1, \ldots, x_L]$ is the matrix of sparse source amplitudes. Assuming the sources are stationary over time, all the columns in $X$ share the same sparsity profile. The observations are assumed to be independent across snapshots giving the multi-snapshot likelihood function

$$p(Y|X) = \prod_{l=1}^L p(y_l|x_l) = \prod_{l=1}^L \mathcal{CN}(y_l; Ax_l, \sigma^2 I).$$  \hspace{1cm} (5)$$

In SBL, the source amplitudes are treated as zero-mean complex Gaussian random vectors with diagonal covariance $\Gamma = $
Fig. 2. Sensor positions for various 2D planar arrays (top row). Estimated 2D spectrum when $K = 25$ sources are present using CBF (second row), MUSIC (third row), and SBL (last row). Circle ‘$\circ$’ indicates true source positions.

The prior distribution is given by

$$p(X) = \prod_{l=1}^{L} p(x_l) = \prod_{l=1}^{L} \mathcal{CN}(x_l; 0, \Gamma).$$  \hfill (6)

From the Gaussian prior (6) and likelihood (5), the evidence term $p(Y)$ is also Gaussian and given by

$$p(Y) = \int p(X)p(Y|X)dX = \prod_{l=1}^{L} \mathcal{CN}(y_l; 0, \Sigma_y),$$  \hfill (7)

where $\Sigma_y = \sigma^2 I + A \Gamma A^H$. In SBL we estimate the diagonal entries of $\Gamma$ by maximizing the (log) evidence i.e. log $p(Y)$

$$\hat{\gamma}_1 \ldots \hat{\gamma}_M = \arg \max_{\gamma} \left\{ - \sum_{l=1}^{L} (y_l^H \Sigma_y^{-1} y_l + \log |\Sigma_y|) \right\}. \hfill$$

Differentiating the above objective function and equating the derivatives to zero gives the fixed point update rule (for details see [15, 16, 18, 21])

$$\hat{\gamma}_m^{\text{new}} = \hat{\gamma}_m^{\text{old}} \frac{\text{Tr}[S_y \Sigma_y^{-1} a_m a_m^H \Sigma_y^{-1}]}{\text{Tr}^{-1}[\Sigma_y^{-1} a_m a_m^H \Sigma_y^{-1}]} \hfill (8)$$

where $S_y = 1/T YY^H$ is the sample covariance matrix (SCM) and $\text{Tr}[]$ the trace operator. The SBL pseudocode is given in Table 1. The noise variance is also required to be estimated [17, 18], here we assume noise to be known. The parameters $\epsilon$ and $N_t$ are the convergence error threshold and number of iterations of the algorithm. The unknown vector $\gamma$ is initialized to the conventional beamformer (CBF) output.

As discussed in [15], estimate of $\gamma$ by SBL is sparse. Since the element $\gamma_m$ of $\gamma$ corresponds to the variance of $x_m$, when $\gamma_m = 0$ it implies that $x_m = 0$. Hence a sparse $\gamma$ corresponds to sparse source amplitude vector $x$. The SBL algorithm in Table 1 when applied to sparse array observations can identify more sources than the number of sensors [7, 8].
Localization accuracy (%)
Co-prime
0 5 10 15 20 25 30
SNR (dB)
0 25 50 75 100
Nested

Fig. 4. 2D Beamforming: 2D Spectrum for MUSIC (left) and SBL (right) as the number of snapshots is increased as \( L = 10, 50, 100, \& 500 \). The array is a URA with 100 sensors.

does not exist as the number of sensors (24) is less than the number of sources (25). SBL provides best resolution for all the three arrays (URA, co-prime, and nested). It can accurately perform localization even when number of sources exceeds the number of sensors.

For various algorithms, percentage of the total sources localized (averaged over 50 Monte Carlo simulations) for \( K = 25 \) sources is shown in Fig. 3. A source is said to be correctly localized if there is a peak in the spectrum within 3° radius of true source location. The localization accuracy is plotted as a function of SNR (Fig. 3, top) and as a function of number of snapshots (Fig. 3, bottom). SBL performs significantly better than CBF for all arrays. SBL performs better than MUSIC for the URA especially for low SNR and low snapshot scenarios. The superior resolution of SBL when compared to MUSIC is also seen when two DOAs are nearby and increasingly more snapshots are used for processing (Fig. 4). In fact for the URA just 10 snapshots with SNR of 15 dB give accurate localization with SBL while MUSIC performs poorly because of snapshot deficiency. In future work we will include experimental data results and comparison with co-array MUSIC [3] which can identify more sources than sensors.

5. CONCLUSIONS

We provided an introduction to 2D co-prime and nested arrays for 2D beamforming. SBL is advantageous since it can directly work with sparse array data. As observed in simulations, compared to CBF and MUSIC, SBL has higher accuracy of DOA localization while using fewer snapshots and in low SNR scenarios. Also, SBL combined with sparse arrays can identify more sources than the number of sensors.

4. SIMULATIONS

We compare 2D beamforming results of SBL with conventional beamforming (CBF) [24] and MUSIC [25, 26]. Three arrays with similar aperture are compared (URA, co-prime, and nested) as shown in Fig. 2 (top row) with basic array spacing of \( \lambda/2 \). The URA consists of \( 10 \times 10 \) sensors while co-prime and nested arrays have \( 9+16-1=24 \) sensors each. We look for sources every 1° in elevation \([0, 90]°\) and azimuth \([0, 360]°\), giving \( 91 \times 360 = 32760 \) possible source locations. \( K = 25 \) sources are generated with uniformly random azimuth and elevation directions and each with source variance \( \gamma = 1 \). A total of \( L = 100 \) snapshots each with an array SNR of 20 dB are simulated. The normalized spectrum obtained by CBF (second row), MUSIC (third row), and SBL (last row) are shown in Fig. 2 for a random data realization. All three methods work directly on the observed SCM \( \mathbf{S}_y \).

For the URA, CBF can distinguish between relatively distant sources but often clusters nearby sources together. MUSIC provides super-resolution with the URA as there are 25 sources and 100 sensors. For co-prime and nested arrays the CBF spectrum is non-informative and the MUSIC spectrum does not exist as the number of sensors (24) is less than the number of sources (25). SBL provides best resolution for all the three arrays (URA, co-prime, and nested). It can accurately perform localization even when number of sources exceeds the number of sensors.

For various algorithms, percentage of the total sources localized (averaged over 50 Monte Carlo simulations) for \( K = 25 \) sources is shown in Fig. 3. A source is said to be correctly localized if there is a peak in the spectrum within 3° radius of true source location. The localization accuracy is plotted as a function of SNR (Fig. 3, top) and as a function of number of snapshots (Fig. 3, bottom). SBL performs significantly better than CBF for all arrays. SBL performs better than MUSIC for the URA especially for low SNR and low snapshot scenarios. The superior resolution of SBL when compared to MUSIC is also seen when two DOAs are nearby and increasingly more snapshots are used for processing (Fig. 4). In fact for the URA just 10 snapshots with SNR of 15 dB give accurate localization with SBL while MUSIC performs poorly because of snapshot deficiency. In future work we will include experimental data results and comparison with co-array MUSIC [3] which can identify more sources than sensors.

Fig. 3. 2D Beamforming: Percentage of sources localized over 100 Monte Carlo simulations for changing SNR (top) and changing snapshot number (bottom). All the arrays have same aperture, number of sources is \( K = 25 \).
6. REFERENCES


